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ERRORS OF UNIVERSAL CONTACTLESS CONVERTERS OF MONITORING AND CONTROL SYSTEMS FROM EXTERNAL MAGNETIC FIELDS

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Abstract: In communication and communication devices, power equipment, relay protection and automation terminals, in the electric power industry of "smart" cities and homes, in industry, in railway transport, microprocessor-based relay protection and automation devices, distributed generation installations, including renewable energy sources, and electricity storage, as well as "intelligent" automated information and measurement systems are beginning to be used. Contactless converters of direct and alternating currents of control and control systems are widely used in them. Their disadvantages are a narrow range of controlled currents, large dimensions and weight. Therefore, it is important to eliminate them. The paper discusses the general principles of construction of contactless converters of large direct currents, the main requirements for them, and shows the results of the development of one of the options proposed by us, universal contactless magneto-modulation converters of large direct currents with an extended range for various control and control systems. They differ from the known ones by an extended controlled range with small dimensions and weight, and increased accuracy and sensitivity. The converter has a simple and technological design with low material consumption and cost, and can control large direct currents, as well as alternating currents, without contact. The paper considers the errors from external magnetic fields of universal contactless converters of control and control systems. It is shown that the error from the external magnetic field does not exceed 0.08% if the number of sections of the measuring winding is even and with their symmetrical arrangement, and with their even increase-the error decreases. At the same time, the developed contactless converters can be widely used in industry, metallurgy, railway transport, agriculture, water and farming, as well as in the electric power industry of "smart" cities and homes and for checking electric meters at the place of their installation.

Key words: non-contact measurement, magnetic modulation converter, meter verification, renewable energy sources, laser installation, split magnetic circuit, integrating circuit, distributed magnetic parameters.

Аннотация: Алоқа ва алоқа қурилмаларида, энергетик усқуналарда, реле ҳимояси ва автоматлаштириш терминалларида, ақли шаҳарлар ва уйларнинг электр энергетикасида, саноатда, темир йўл транспортида микропроцессорга асосланган реле ҳимояси ва автоматлаштириш қурилмалари, қайта тикланадиган энергия манбалари, шу жумладан, тақсимланган генерация қурилмалари, электр энергияси, шунингдек "интеллектуал" автоматлаштирилган ахборот ва ўлчаиш тизимлари қўлланила бошланди. Уларда кузатув ва бошқариш тизимларининг тўғридан-тўғри ва ўзгарувчан тоқларининг контактсиз конверторлари кенг қўлланилади. Уларнинг камчиликлари тор доирадаги бошқариладиган тоқлар, катта ўлчамлар ва оғирликдир. Шунинг учун уларни бартараф қилиш муҳим саналади. Ушбу мақолада контактсиз юқори доимий ток конверторларини тизилишининг умумий принциплари, уларга қўйиладиган асосий талаблар кўриб чиқилган ва биз таклиф қилган вариантлардан бирини ишлаб чиқиш натижалари кўрсатилган, ҳар хил кузатиш ва бошқариш тизимлари учун кенгайтирилган диапазонга эга универсал доимий магнит-модуляцияли юқори тоқли конверторлар таклиф этилган. Улар маълум бўлганларидан кичик ўлчамлари ва вазни, яхшиланган аниқлиги ва сезгирлиги, кенгайтирилган бошқариладиган диапазон билан фарқланади. Конвертер оддий ва технологик дизайнга эга бўлиб, материал кам сарфланади ва арзон таннархга эга ҳамда бевосита катта оқимларни, шунингдек ўзгарувчан тоқларни контактсиз бошқариши мумкин. Мақолада ташиқи магнит майдонларини назорат қилиш ва бошқариш тизимларининг универсал контактсиз конверторларини хатоликларини кўриб чиқилган. Агар ўлчаиш ўрамлири

секторининг сони жуфт ва улар носимметрик жойлашган бўлса, ташиқи магнит майдонининг хатолиги 0,08 % дан ошмаслиги ва уларнинг сони тоқ бўлса, хатolik камайиши кўрсатилган. Шу билан биргаликда, ишлаб чиқарилган контактсиз конверторлар саноат, металлургия, темир йўл транспорти, қишлоқ хўжалиги, сув ва фермер хўжалигида, шунингдек, ақлли шаҳарлар ва уйларнинг электр энергетикасида ва электр ҳисоблагичларни ўрнатиши жойларида қийёслашда кенг қўлланилиши мумкин.

Таянч сўзлар: контактсиз ўлчаши, магнит-модуляцияли ўзгарткич, ҳисоблагичларни қийёлаши, қайта тикланадиган энергия манбалари, лазер қурилмаси, бўлинадиган магнитўтказкич, интегралловчи контур, тақсимланган магнит параметрлари.

Аннотация. В устройствах связи и коммуникации, силовом оборудовании, терминалах релейных защит и автоматики, в электроэнергетике “умных” городов и домов, в промышленности, на железнодорожном транспорте начинают применяться микропроцессорные устройства релейной защиты и автоматики, установки распределенной генерации, включая возобновляемые источники энергии, и накопители электроэнергии, а также «интеллектуальные» автоматизированные информационно-измерительные системы. В них широко применяются бесконтактные преобразователи постоянного и переменного токов систем контроля и управления. Их недостатки – узкий диапазон контролируемых токов, большие габариты и масса. Поэтому важным является их устранение. В работе рассмотрены общие принципы построения бесконтактных преобразователей больших постоянных токов, основные требования к ним и показаны результаты разработки одного из вариантов, предложенных нами, универсальных бесконтактных магнитомодуляционных преобразователей больших постоянных токов с расширенным диапазоном для различных систем контроля и управления. Они отличаются от известных расширенным контролируемым диапазоном при малых габаритах и массе и повышенными точностью и чувствительностью. Преобразователь имеет простую и технологичную конструкцию при низких их материалоемкости и стоимости и может бесконтактно контролировать большие постоянные токи, а также и переменные токи. В работе рассмотрены погрешности от внешних магнитных полей универсальных бесконтактных преобразователей систем контроля и управления. Показано, что погрешность от внешнего магнитного поля не превышает 0,08%, если количество секций измерительной обмотки будет четным и при их симметричном расположении, а при их четном увеличении – погрешность уменьшается. При этом разработанные бесконтактные преобразователи могут широко применяться в промышленности, металлургии, на железнодорожном транспорте, в сельском, водном и фермерских хозяйствах, а также и в электроэнергетике “умных” городов и домов и для поверки электрических счетчиков на месте их установки.

Ключевые слова: бесконтактное измерение, магнитомодуляционный преобразователь, поверка счетчиков, возобновляемые источники энергии, лазерная установка, разъемный магнитопровод, интегрирующий контур, распределенные магнитные параметры.

Introduction

The use of automated monitoring and control systems for various technological and physical processes in industry, metallurgy, and the agro-industrial sphere is characterized by the widespread use of primary means of collecting and processing information [1]. At the same time, the primary converter is a link of any information - measuring or control system and almost completely determines its metrological characteristics. Losses in the accuracy and reliability of the final result associated with the unsuccessful use of the primary converter are unable to restore even the most perfect information transformation system. The task becomes much more complicated in case of exposure to unstable factors, such as changes in temperature and humidity of the environment, the influence of corrosive media, electric and magnetic fields, vibrations, radiation, etc.

The need to convert large currents in various sectors of the national economy and, in particular, in the chemical industry, on railway transport, in metallurgy, land reclamation, irrigation and in agriculture in general, arises when monitoring and controlling the operating modes of powerful electric motors, substations and various consumers, where used non-contact measuring ferromagnetic transducers of high currents (FTHC) [4,5].

The need to break the current circuit for the temporary switching on of electrical measuring instruments, the presence of large power losses in the measuring current transformers and shunts, the undesirability or impossibility of breaking the circuit under the conditions of the technological process, as well as the safety requirements, led to the contactless conversion and measurement of large direct and alternating currents in circuits without their break, i.e. without destroying the integrity of the busbar [5,6].

Consideration of the issues of converting large direct currents (LDC) in electric power systems and electrical technological installations showed that one of the reasons for their low efficiency is the unsatisfactory technical characteristics of their secondary systems - control and monitoring systems for the operating modes of electrical power and electrical technological installations and, in particular, used in them FTHC. It is noted that FTHCs used in secondary systems of electric power systems and electrical installations should have an adjustable conversion range, better dynamic properties in transient modes of operation of electrical power systems and electrical installations, as well as stability of characteristics under extreme operating conditions [7 - 28].

When considering the places of non-destructive contactless testing of high currents, the main requirements for FTHC were identified. These include: high accuracy, reliability, sensitivity, low weight, dimensions, material consumption and cost, manufacturability of design, absence of errors from the influence of external magnetic fields, a reverse bus with a current from the center of the integrating circuit, ferromagnetic masses, absence of galvanic coupling between the measured alternating current and the measuring circuit and the presence in some cases of the possibility of both a fixed regulation of the sensitivity of the transducers in a wide range of convertible large alternating currents and the flexibility of the integrating circuit, and the design of the FTHC both portable and stationary [29 - 56].

In this regard, it is very important to develop and study such FTHCs that would have an extended range of convertible high currents with small dimensions and weight and increased accuracy, a simplified and technological design with low material consumption and cost [57, 58].

The solution to this problem can be facilitated by the development of universal magnetomodulation contactless converters (UCC) of control and management systems in industrial and agroelectric power engineering.

Research methods and results obtained

The main tasks that need to be solved in the design of UCCs in accordance with modern trends in the development of measuring instruments and conversions of high currents are mainly expanding the range of measured and converted values of high currents, expanding their frequency range in the case of alternating currents, increasing accuracy and sensitivity, decreasing weight and overall dimensions of UCC [40]. Most often, these tasks must be solved in aggregate for one UCC, choosing circuits and designs that simultaneously satisfy the requirements of a wide range of linearity of the static characteristic, a low threshold of sensitivity and a wide controlled current range with a small volume and low weight - overall characteristics of the UCC. In this regard, it is worthy of special attention to consider ways and methods,

It is shown that one of the effective possibilities of expanding the range of linearity of the UCC and reducing the sensitivity threshold is to increase the length of the split magnetic circuit of the UCC and its cross-sectional area to the maximum permissible size [2].

We have developed a number of UCCs, in which the set tasks are solved by using in the developed converters special designs of detachable closed magnetic circuits with transversely and longitudinally distributed magnetic parameters and an increased length of the working magnetic flux path over steel [2,40]. Next, we will consider the design features of one of the developed UCC and investigate its errors from the influence of external magnetic fields.

A generalized version of the developed UFT is shown partially with the main dimensions in Fig. 1. This design is developed on the basis of the UCC [11] and is a UCC with transversely and longitudinally distributed magnetic parameters. It features increased sensitivity and an extended range of converted currents. UCC contains a closed magnetic circuit, consisting of two groups of trapezoidal ferromagnetic elements 1 and 2. The first group in the upper ring includes identical ferromagnetic elements 1 and the second group - in the lower ring - also includes identical plate-like ferromagnetic elements 2 with the same gaps between them. Ferromagnetic elements 1 and 2 are assembled from separate plates made of thin sheet electrical steel. Each ferromagnetic element has two through holes,

through each of which a modulation winding is wound, consisting of sections 4 and 6. Sections 4 and 6 are connected in series and according to. A measuring winding 5 is wound between the through-holes over the modulation winding. All measuring windings are connected in series with each other and closed to a measuring device, and the modulation windings are also connected in series and connected to a stable AC source (not shown in Fig. 1). In order to freely grip the bus 7 with controlled current, the closed magnetic circuit 1 is made detachable. The series connection of the modulation windings 4 and 6 with each other in the presence of alternating current and the arrangement of the measuring windings 5 in the intervals between the through holes in the ferromagnetic elements allowed Sections 4 and 6 are connected in series and according to. A measuring winding 5 is wound over the modulation winding between the through holes. All measuring windings are connected in series with each other and closed to the measuring device, and the modulation windings are also connected in series and connected to a stable AC source (not shown in Fig. 1). In order to freely grip the bus 7 with controlled current, the closed magnetic circuit 1 is made detachable. The series connection of the modulation windings 4 and 6 with each other in the presence of alternating current and the arrangement of the measuring windings 5 in the intervals between the through holes in the ferromagnetic elements allowed Sections 4 and 6 are connected in series and according to. A measuring winding 5 is wound between the through-holes over the modulation winding. All measuring windings are connected in series with each other and closed to a measuring device, and the modulation windings are also connected in series and connected to a stable AC source (not shown in Fig. 1). In order to freely grip the bus 7 with controlled current, the closed magnetic circuit 1 is made detachable. The series connection of the modulation windings 4 and 6 with each other in the presence of alternating current and the arrangement of the measuring windings 5 in the intervals between the through holes in the ferromagnetic elements allowed All measuring windings are connected in series with each other and closed to the measuring device, and the modulation windings are also connected in series and connected to a stable AC source (not shown in Fig. 1). In order to freely grip the bus 7 with controlled current, the closed magnetic circuit 1 is made detachable. The series connection of the modulation windings 4 and 6 with each other in the presence of alternating current and the arrangement of the measuring windings 5 in the intervals between the through holes in the ferromagnetic elements allowed All measuring windings are connected in series with each other and closed to the measuring device, and the modulation windings are also connected in series and connected to a stable AC source (not shown in Fig. 1). In order to freely grip the bus 7 with controlled current, the closed magnetic circuit 1 is made detachable. The series connection of the modulation windings 4 and 6 with each other in the presence of alternating current and the arrangement of the measuring windings 5 in the intervals between the through holes in the ferromagnetic elements allowed to carry out longitudinal modulation of the magnetic resistance of the magnetic circuit along the path of the working flow Φ , created by a controlled direct current, and induce an EMF in the measuring windings 5, depending on the converted direct current. The developed DCL can also control alternating current. In this case, there should be no alternating current in sections 4 and 6 of the modulation winding.

The expansion of the upper limit of the controlled direct current in the developed DCC design is carried out by increasing the length of the working magnetic flux along the steel of the magnetic circuit elements and including transverse and longitudinal air gaps in its path, i.e., making a split magnetic circuit with transversely and longitudinally distributed magnetic parameters.

To control the BPT with a detachable magnetic circuit, the UCC cover the bus 7. Due to the modulation ampere-turns, the split magnetic circuit is in a saturated state during each half-period of the supply voltage. In this case, the permeability of the magnetic circuit for the longitudinal field created by the controlled current decreases sharply. At the moment when the modulation current passes through zero, the magnetic core permeability rises to the initial value. Thus, with the stability of the modulation ampere turns, an EMF of double frequency will be induced in the measuring winding, depending on the controlled current.

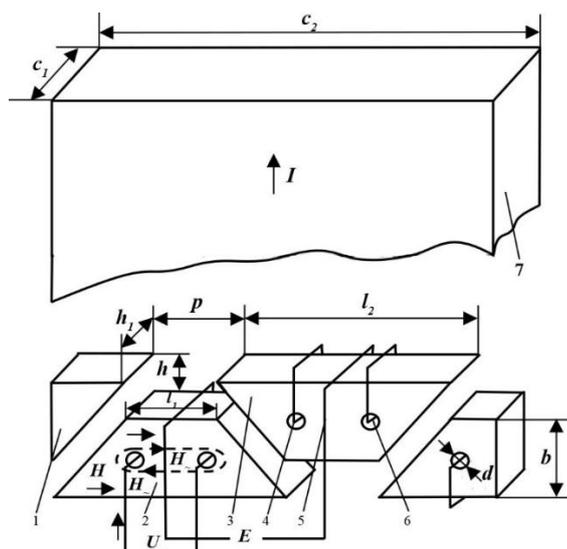


Fig. 1. Part of a universal non-contact magneto-modulation converter for large direct currents.

With the mutual movement of halves 2 and 3 of the split magnetic circuit of the UCC, the size of the gaps between the trapezoids changes, leading to a change in the whole of the magnetic resistance of the magnetic circuit in the path of the working magnetic flux Φ created by a controlled direct current. This leads to a change in the limits of the controlled current, i.e. allows you to make the UCC multi-limit.

For widespread use of the developed DCS in multidisciplinary monitoring and control systems, they must be free from the influence of external magnetic fields. Therefore, let us investigate the influence of external magnetic fields on the developed CCA.

In the presence of an external magnetic field, an additional error appears UTP caused by induction of EMF in the measuring winding from the tangential component of the external magnetic field strength.

In fig. 2 shows the outline UTP in an external uniform magnetic field in the absence of a magnetic circuit in it. The tangent component of the external magnetic field strength $H_{\theta H}$ at any point of the contour is

$$H_{T\theta H} = H_{\theta H} \sin \varphi, \quad (1)$$

where φ is the central corner.

Consideration of the influence of the magnetic circuit UTP is carried out using the design parameter P_k . Then the expression for the tangential component of the external magnetic field strength at any point detachable magnetic circuit UCC can write

$$H_{T\theta H\mu} = H_{T\theta H} \cdot P_k = H_{\theta H} P_k \sin \varphi. \quad (2)$$

Multiplying both parts of (2) by the coefficient of approximation of the magnetization curve a_2 , we obtain the dimensionless value of the tangent component of the external magnetic field strength at any point detachable magnetic circuit UCC as

$$a_2 H_{T\theta H\mu} = H_{x\theta H} \sin \varphi, \quad (3)$$

where

$$H_{x\theta H} = a_2 H_{\text{vnpk}} \dots \quad (4)$$

The reduced additive error due to the influence of an external magnetic field with an approximation sufficient for analysis is determined from the expression

$$\Delta_{\theta H A} = \frac{\Delta E_{\Sigma \theta H}}{E_{\max}} \dots \quad (5)$$

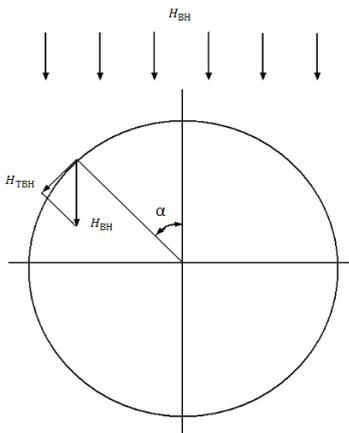


Fig. 2. Circuit UTP in an external uniform magnetic field in the absence of a magnetic circuit in it.

Static characteristic UTP is

$$E_{cp} = \frac{\omega a_1 w_u S}{\pi} [2arctgH_x - arctg(H_x - 1,65) - arctg(H_x + 1,65)] \tag{6}$$

or

$$E_{cp} = \frac{E_{\sigma}}{\pi} [2arctgH_x - arctg(H_x - 1,65) - arctg(H_x + 1,65)]. \tag{7}$$

Here E_b is the base value of the output EMF, equal to

$$E_{\sigma} = \omega a_1 w_u S, \tag{8}$$

H_x - the strength of the magnetic field penetrating the turns of the measuring winding, which is a function of the controlled current at a given value of the modulation current and carries information about the design parameters of the split closed magnetic circuit and the grade of steel used in the magnetic circuit.

Substituting instead of H_x in (7) the value (3) and dividing the result by 2π , we obtain the change in the EMF from the external magnetic field at each point of the circuit integration provided that the entire measuring winding is evenly distributed over contour in the form

$$\Delta e_{\varphi BH} = \frac{E_{\sigma}}{2\pi^2} [2arctgH_{x_{\theta H}} \sin \varphi - arctg(H_{x_{\theta H}} - 1,65) - arctg(H_{x_{\theta H}} + 1,65)]. \tag{9}$$

Here $\frac{E_{\sigma}}{2\pi}$ - the base value of the output EMF for each angular radian of the DCC integrating circuit.

To simplify the calculation of the obtained expression, let us take each section of the measuring winding concentrated at a point. Then the central angle of each measurement point in the integrating circuit of the DCS is equal to

$$\varphi = i2\pi / m, \tag{10}$$

where m is the number of sections of the measuring winding; $i = 1, 2, \dots, m$.

In this case, the average EMF of the DCS is equal to

$$\Delta E_{\Sigma \theta H} = \frac{E_{\sigma}}{\pi m} [2arctgH_{x_{BH}} \sin i \frac{2\pi}{m} - arctg(H_{x_{BH}} \sin i \frac{2\pi}{m} - 1,65) - arctg(H_{x_{BH}} \sin i \frac{2\pi}{m} + 1,65)]. \tag{11}$$

The maximum value of the output EMF of the DCC at $H_x = 0.7$, in turn, is

$$E_{max} = 0,65E_b. \tag{12}$$

Substituting (7) and (11) into (5) and taking into account (12), we obtain the reduced additive error from the influence of an external magnetic field

$$\Delta E_{\theta H A} = \frac{0,53}{m} \sum_{i=1}^m [2arctgH_{x_{BH}} \sin i \frac{2\pi}{m} - arctg(H_{x_{BH}} \sin i \frac{2\pi}{m} - 1,65) - arctg(H_{x_{BH}} \sin i \frac{2\pi}{m} + 1,65)] \tag{13}$$

By expression (13), the reduced additive error from the influence of an external magnetic field was determined at $m = 6; 12$ and 18 . The calculation results showed that $\Delta_{\text{внА}}$ does not exceed 0.06% at $m = 6$ and $H_{\text{вн}} = 1.2$. With an even increase in m , the error decreases.

Experiments conducted with the UTP, the parameters of which (see Fig.1): the outer diameter of the split closed magnetic circuit $D_{\text{внвн}} = 0.205$ m; internal diameter of the split closed magnetic circuit $D_{\text{внвнм}} = 0.175$ m; $h_1 = 2.8 \cdot 10^{-3}$ m; $h = 1 \cdot 10^{-3}$ m; $p = 2 \cdot 10^{-3}$ m; $b = 15 \cdot 10^{-3}$ m; horizontal length of the inclined part of the ferromagnetic elements $X_m = 17.9 \cdot 10^{-3}$ m; the number of ferromagnetic elements $n_{\Sigma} = 30$, and the split magnetic core is made of cold-rolled steel 3414 with a sheet thickness of 0.35 mm, for which the approximation coefficients are $a_1 = 1.853$ T and $a_2 = 0.714 \cdot 10^{-2}$ m/A, showed that the reduced additive error from the influence of an external magnetic field does not exceed 0.08% , if the number of sections of the measuring winding is even $t = 2k$, where $k = 1, 2, \dots, \frac{m}{2}$ and with their symmetrical arrangement.

Conclusion

Developed universal non-contact wide-range magnetomodulation converters of large direct and also alternating currents for modern control and management systems in metallurgy, railway transport, solar and laser technology, renewable energy sources, industry, agro-industrial sphere, as well as for verification of electric meters at the place of their installation, characterized by an extended controlled range of convertible direct currents with small dimensions and weight, increased accuracy and sensitivity, simplicity and manufacturability of the design with low material consumption and cost and the possibility of contactless control of direct and alternating currents with an error 1.5% , as well as for the control of electricity and verification of electricity meters at the place of their installation. The errors of universal contactless converters of monitoring and control systems from external magnetic fields are considered. The results of their research are presented. Shown that the error from the external magnetic field does not exceed 0.08% if the number of sections of the measuring winding is even and with their symmetrical arrangement, and when they increase evenly, the error decreases.

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